

Freyja: A Full Multirotor System for Agile & Precise Outdoor Flights

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Abstract—Several independent approaches exist for state estimation and control of multirotor unmanned aerial systems (UASs) that address specific and constrained operational conditions. This work presents a complete end-to-end pipeline that enables precise, aggressive and agile maneuvers for multirotor UASs under real and challenging outdoor environments. We leverage state-of-the-art optimal methods from the literature for trajectory planning and control, such that designing and executing dynamic paths is fast, robust and easy to customize for a particular application. The complete pipeline, built entirely using commercially available components, is made open-source and fully documented to facilitate adoption. We demonstrate its performance in a variety of operational settings, such as hovering at a spot under dynamic wind speeds of up to 5–6 m/s (12–15 mi/h) while staying within 12 cm of 3D error. We also characterize its capabilities in flying high-speed trajectories outdoors, and enabling fast aerial docking with a moving target with planning and interception occurring in under 8 s.

I. INTRODUCTION

Field applications of multirotor unmanned aerial systems (UASs) have become increasingly realistic and far-reaching over the last decade. This is due, in part, to a sustained development of their potential as field agents that work in real and complex environments found ‘in the wild’. Modern use-cases for multirotors span the breadth of environmental sciences (profiling the lower atmosphere [1], monitoring soil and crops [2], studying water bodies [3], etc), and autonomous search and rescue operations [4]. While these have advanced the capabilities of multirotors, they do not always require precise and accurate control of the trajectories of the multirotor. The next generation of outdoor applications, such as intercepting objects in the air [5] and docking with moving aircraft [6] will require significant advances in state estimation and control implementations, demonstrated outdoors.

To realize such agile, precise and interactive field missions, we must account for natural and loosely modeled phenomena (such as wind and aerodynamic drag), and deviations from expected model parameters (such as the total mass, changing battery voltage, idealized transfer functions etc.) that pose challenges for accurate flights. These adversely affect the performance of a controller, and are more noticeably evident when flying complex time-bound trajectories. Robust compensation for such dynamic effects typically require either extremely customized solutions, or are limited to more constrained and simulated indoor/lab settings. At present, there is a gap between the research/prototype state-of-the-art approaches [7], [8], [9], and their full realization as field

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Fig. 1: Snapshots depicting instances of a multirotor UAS in different outdoors scenarios: (top) intercepting parachutes mid-air, (bottom) flying aggressive circles around a spot.

agents. We are currently lacking a complete and generalized end-to-end pipeline for high-level state estimation and precise control over aggressive trajectories outdoors.

In this paper, we introduce such a pipeline that we call *Freyja*, that addresses this gap through efficient, modular elements that fit together cohesively on small onboard computers. We position this work in the context of systems and components that are cost effective, commercially available, and require no specific customizations. By building on a modular architecture using robust and individually optimal elements, we show a complete system that can not only measure and reject unexpected extrinsic disturbances found in field missions, but also extend the envelope of such missions by performing precise, aggressive and feedforward maneuvers usually confined indoors. Figure 1 depicts two instances of such missions where a multirotor is required to exercise precise control for intercepting airborne parachutes, and for flying aggressive trajectories outdoors.

The system presented in this work is designed around a small-sized quadcopter frame equipped with an attitude-stabilizing autopilot (such as the popular Pixhawk). Our approach builds around three key enablers that address localization, trajectory formulation and control. For localization, we use a miniaturized low-power real-time kinematic (RTK) GPS unit for precise global and map-frame positioning. This data, fused with inertial measurements through an Extended Kalman Filter (EKF), provides the fast and accurate system state required by a controller. We allow a wide scope for trajectories, ranging from discrete waypoints and discontinuous paths, to continuous and smooth parametric curves.

The control strategy utilizes a linear quadratic gaussian (LQG) control (which is a tandem implementation of a linear quadratic regulator (LQR) and a full-state Kalman filter) [7] along with trajectory feed-forward components to precisely track a reference trajectory in time and space. The observer in LQG is capable of measuring 3-axis extrinsic disturbances acting upon the system, which allows the feedback controller to reject them in the successive iterations. The system is feedback linearized over a nested autopilot loop, exploits the differential flatness of a multirotor system, and uses a non-linear inversion map to generate control inputs to the autopilot. This allows highly dynamic trajectories (and their feed-forward components) to be planned entirely in the output space using any of the classical planning methods. The proposed system remains oblivious to the type of multirotor (quad-, hexa- etc) by delegating the low-level attitude stabilization to a well-tuned autopilot.

The key contributions of this work are:

- A complete end-to-end pipeline that addresses state estimation, trajectory generation, and precise control under challenging outdoor conditions;
- An analysis of the impact of developing feedforward control & optimal bias observers for real environments;
- Outdoor evaluations and demonstrations of trajectory control for translational speeds over 6 m/s, hovering with a 3D error of less than 4 cm, and precise control for aerial docking with a moving target in under 8 s.

II. BACKGROUND

Fast and accurate estimates of the inertial position and velocity of the UAS in outdoor environments is key to precise trajectory control. The requirements in precision may vary for different applications; an initially coarse estimate might suffice for large-area applications such as search and rescue [4], [10]. An extremely high precision, on the order of a few centimeters, is necessary for closer interactions such as inspecting structures [11], landing on targets [12], or perching on power lines [13]. Consumer-grade global positioning systems are severely restrictive in such cases, with stated accuracies well above 1.5 m [14]. Consequently, several of these applications fuse visual-inertial data from onboard cameras and lasers. When GPS is available, differential solutions and real time kinematic (RTK) systems can offer significantly higher accuracies (on the order of 2–3 cm). Fusing low-rate RTK data with IMU measurements and/or visual odometry (VIO) has shown highly promising results [15], [16]. This is enabled by newer commercially available solutions that are miniaturized enough to be retrofitted to small multirotors.

Several state feedback and control approaches have been also developed for underactuated systems (for instance, [7] and references therein). For multirotors, these are developed using system model representations that are extremely detailed [8] or more abstract [17], depending on the context of the problem. Indoors, and in semi-structured environments, where motion-capture or VIO can provide reliable state information, multirotors have been used to demonstrate agile maneuvering tasks [18], grasping objects [19], and agile

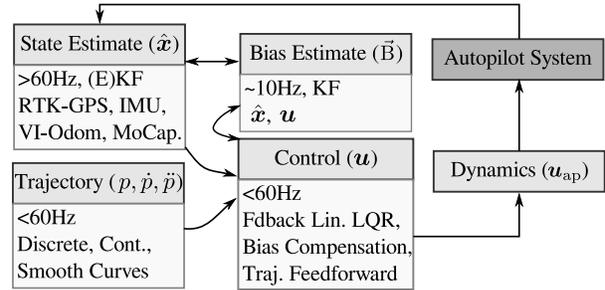


Fig. 2: A block diagram representation of the system architecture. We address each of the modules independently, and make them amenable to drop-in replacements.

load transport [9]. While some of these approaches may be transferable to systems ‘in the wild’, we still lack detailed evaluations outdoors.

Our approach here is developed using a similar high-level (point mass) representation that encapsulates nested autopilot loops so that the resultant system can generalize better. Complex system models that account for aerodynamic effects such as blade flapping and aerodynamic drag can be crucial for aggressive flight regimes [15], [20], however, their application to outdoor flight has been fairly limited. Similarly, trajectory generation methods that exploit a UAS’s differential flatness and shape smooth accelerations have been demonstrated [18] primarily for constrained indoor environments. Recent work has demonstrated such methods outdoors applied to aerial docking missions [21]. Our objective is to bridge this gap with a complete system that can perform agile maneuvers outdoors under real disturbances.

III. TECHNICAL DETAILS

Figure 2 shows a block-diagram view of our architecture, where each shaded rectangle represents a modular component of the complete pipeline. We will describe the individual modules in a logical progression in the following subsections. Note that each module is capable of having drop-in replacements in the form of alternative choices of sensors, control system and planning.

We let \mathcal{W} represent the world-fixed NED (north-east-down) coordinate frame. In the following text, a local (map) frame, \mathcal{M} , is assumed to be rigidly fixed in \mathcal{W} , with its axes aligned with \mathcal{W} and its origin initialized where the UAS is initialized. The translational position, $P^{\mathcal{M}}$, and the velocity, $\dot{P}^{\mathcal{M}}$, of the UAS are expressed in this local frame. We assume that the rotation angles and the rates, both expressed in the vehicle’s body frame, are handled by the autopilot.

A. System Model

We develop the estimation and control pipeline on a feedback-linearized translational system model of the UAS, incorporating elements from classical approaches in literature [8], [17]. A distinguishing element in our design is the separation of the controller state from the observer state. The model is derived from the dynamics of a rigid body system (b) with six degrees of freedom (DOF) with mass m ,

$$m\ddot{\mathbf{a}} = -R_{\mathcal{M}}^b \cdot T + \hat{e}_d mg, \quad (1)$$

175 where $R_a^b \in \text{SO}(3)$ denotes the 3×3 rotation between the
 176 frames a and b , T is the collective thrust produced by the
 177 rotors, g is the acceleration due to gravity and \hat{e}_d denotes
 178 a unit vector along the vertical (down) axis of the inertial
 179 frame. The matrix $R_{\mathcal{M}}^b$ is obtained from the Euler roll (ϕ),
 180 pitch (θ) and yaw (ψ) angles of the UAS body in the Z-
 181 Y-X rotation order. Thus, by assuming that desired values
 182 of these angles and a collective thrust can be maintained by
 183 an autopilot’s “inner loop”, we can affect a desired linear
 184 acceleration, $\vec{a} \in \mathbb{R}^3$, of the body in the inertial frame. We
 185 therefore define the control command sent to the autopilot as
 186 $\mathbf{u}_{\text{ap}} = [\phi_d, \theta_d, \psi_d, T_d]^\top$ composed of the desired values of
 187 these quantities.

The non-linear system defined by Eqn (1) lets us model a
 linear system with second-order dynamics with accelerations,
 \vec{a} , as its inputs. For this system, we define a state vector,

$$\mathbf{x} \equiv [P^{\mathcal{M}}, \dot{P}^{\mathcal{M}}, \psi]^\top \\ = [p_n, p_e, p_d, v_n, v_e, v_d, \psi]^\top, \quad (2)$$

188 composed of the translational position, velocity and the
 189 heading of the UAS, all expressed in the inertial frame. The
 190 dynamics can then be expressed in the traditional form,

$$\dot{\mathbf{x}} = A\mathbf{x} + B\mathbf{u}, \text{ and, } y = C\mathbf{x}, \quad (3)$$

191 with,

$$A = \begin{pmatrix} 0_{3 \times 3} & \mathbb{I}_{3 \times 3} & 0_{3 \times 1} \\ 0_{3 \times 3} & 0_{3 \times 3} & 0_{3 \times 1} \\ 0_{1 \times 3} & 0_{1 \times 3} & 0 \end{pmatrix}, B = \begin{pmatrix} 0_{3 \times 3} & 0_{3 \times 1} \\ \mathbb{I}_{3 \times 3} & 0_{3 \times 1} \\ 0_{1 \times 3} & 1 \end{pmatrix}, C = I.$$

192 The control input to this feedback-linearized system is a
 193 4-vector composed of the translational accelerations from
 194 Eqn (1) and a body-frame rotational rate, $\dot{\psi}$, such that,

$$\mathbf{u} \equiv [\vec{a}, \dot{\psi}]^\top. \quad (4)$$

195 Thus, if appropriate acceleration control inputs, \mathbf{u} , are known
 196 for the linearized system, we can decompose them into \mathbf{u}_{ap}
 197 by a non-linear inversion of Eqn (1).

198 B. State Estimation

199 We generally require a robust and reliable source of state
 200 information to perform accurate and high-speed maneuvers.
 201 To prevent erroneous feedback control, we further require
 202 this information to be updated faster than the control cycle.
 203 Typical GPS systems offer update rates that are too low
 204 (≈ 10 Hz) and are often too inaccurate. For instance, a high-
 205 end GPS accuracy of 0.8 m can be almost twice the diameter
 206 of medium-sized multirotors. For localized operations (within
 207 a radius of 1–2 km), we therefore switch to ground-based
 208 augmentation systems (GBAS) to achieve significantly higher
 209 accuracy in measurements. This is realized in the form of real
 210 time kinematic (RTK) GPS systems that can produce position
 211 measurements with more than 5 cm of accuracy at a similar
 212 rate. The accuracy also remains fairly consistent within the
 213 operational range of RTK systems.

214 We split the state estimation into two separate “processes”
 215 – one that estimates the controllable system states defined in
 216 model, and another that estimates a state model with biases.

217 An optimal state estimator for both allows a controller to
 218 optimally regulate the state by *certainty equivalence*. By the
 219 separation principle, we also know the combined system will
 220 retain its stability guarantees. This also lets us design these
 221 modules independently.

Controller States. For agile maneuvering, RTK-GPS data is
 fused with inertial measurements from an onboard IMU (in
 the autopilot). We adopt an Extended Kalman filter (EKF)
 formulation, and rewrite the non-linear system as

$$\dot{\mathbf{x}} = f(\mathbf{x}, \mathbf{u}, u), \\ z_{\text{pos}} = h_1(\mathbf{x}, v), z_{\text{imu}} = h_2(\mathbf{x}_b, w) \quad (5)$$

222 where f, h_1 and h_2 represent the state transition and mea-
 223 surement maps, \mathbf{x}_b is a new state variable containing only
 224 the attitude angles in the body frame, and u, v, w are the
 225 corresponding zero-mean additive noises over a Gaussian
 226 distribution. The filter then estimates $\hat{\mathbf{x}}$ at a sufficiently high
 227 rate for the controller. The product of this block, eventually,
 228 is the best estimate of the state, $\hat{\mathbf{x}}$, as defined above and
 229 expressed in \mathcal{M} . Several other fusion methods, such as
 230 visual-inertial odometry (VIO), and visual pose estimation
 231 from onboard cameras [22], [23] or motion-capture systems
 232 could provide the state information at a sufficiently high rate.

Observer States. To design the state observer in LQG, we
 augment the state vector in Eqn (2) to include extrinsic
 time-varying forces. We represent these in the form of
 accelerations acting upon the system, so that for the bias
 observer, the augmented system model is represented by

$$\mathbf{x}_B \equiv [\mathbf{x}^\top, \vec{\mathbf{B}}^\top]^\top \quad (6)$$

$$\dot{\mathbf{x}}_B = A_B \mathbf{x}_B + B_B \mathbf{u}, \text{ and, } y_B = C \mathbf{x}_B \quad (7)$$

$$\text{with, } A_B = \begin{pmatrix} A & \mathbb{I}_{3 \times 3} \\ 0_{3 \times 7} & 0_{3 \times 3} \end{pmatrix} \text{ and } B_B = \begin{pmatrix} B \\ 0_{3 \times 4} \end{pmatrix}$$

233 such that, $\vec{\mathbf{B}} = [b_n, b_e, b_d]^\top$ denotes the 3-axis external
 234 disturbances that act as biases on the system.

235 In aggressive maneuvering, aerodynamic drag plays a
 236 significant role in the dynamics [15], [20]. Instead of ex-
 237 plicitly modeling it, we let the bias estimator measure it as
 238 an external force, which a controller can then compensate
 239 for. By appropriate pole-placement of the estimator, the
 240 dynamics of the estimator can be fast enough to measure
 241 other deviations from the system model such as an incorrect
 242 mass (m) variable, an off-center loading, or a changing thrust
 243 due to battery voltage.

244 C. Control

245 The control input, \mathbf{u} , from Eqn (4) applied to the system
 246 is designed with three components, such that,

$$\mathbf{u} \equiv \mathbf{u}_{\text{fb}} + \mathbf{u}_{\text{bc}} + \mathbf{u}_{\text{ff}}, \quad (8)$$

247 where the subscripts fb, bc and ff denote the feedback, bias
 248 compensation, and the feed-forward elements of the signal.
 249 Similar feedforward designs based on differential flatness of
 250 the multirotor system have been employed previously [17].
 251 For outdoor flights where external disturbances can manifest
 252 in several time-varying forms, the bias compensation term

253 plays a very significant role. Our modeling of these distur- 302
 254 bances as accelerations let us incorporate corrections directly 303
 255 into the the control equation.

256 **Feedback.** For a linear system model described by Eqs (2)- 304
 257 (3), it is possible to design a feedback control law that regu- 305
 258 lates the state vector, \mathbf{x} , and drives the error exponentially 306
 259 to zero. Denoting a reference state in time as \mathbf{x}_r , we write 307
 260 the feedback control equation as 308

$$\mathbf{u}_{fb} = -K(\mathbf{x} - \mathbf{x}_r), \quad (9)$$

261 where K is the feedback gain matrix. Substituting \mathbf{u}_{fb} for \mathbf{u} 312
 262 in Eqn (3), the resultant system dynamics can be rewritten as 313
 263 $\dot{\mathbf{x}} = (A - BK)\mathbf{x} = \tilde{A}\mathbf{x}$. For a stable system, the eigenvalues 314
 264 of \tilde{A} must lie strictly on the left-half of the complex plane. 315
 265 Thus, the design matrix K can be chosen to affect a desired 316
 266 pole placement for the system.

267 Theoretically, this feedback gain matrix can be chosen 318
 268 to produce an arbitrarily fast convergence to the desired 319
 269 \mathbf{x}_r . In practice, physical constraints on the system (such as 320
 270 motor response time, clipped battery power, etc) limit large 321
 271 changes in the control effort between successive time steps. 322
 272 Furthermore, a smoother control is often more desirable in 323
 273 many practical applications such as environmental sensing 324
 274 and interactions. Thus, we use a Linear Quadratic Regulator 325
 275 (LQR) design to select an optimal feedback gain matrix K 326
 276 that balances the control expenditure of the system against its 327
 277 ability to regulate state errors. This feedback matrix, denoted 328
 278 K_{lqr} , is the solution for an Algebraic Riccati Equation (ARE) 329
 279 that minimizes the cost functional 330

$$J(\mathbf{x}, \mathbf{u}) = \int_0^\infty \mathbf{x}_e^T Q \mathbf{x}_e dt + \int_0^\infty \mathbf{u}_{fb}^T R \mathbf{u}_{fb} dt.$$

281 **Bias Compensation.** Recall from Section III-B that the state 331
 282 observer models external disturbances acting on the UAS 332
 283 as accelerations (or, equivalently as forces) in the three 333
 284 translational axes. Since the control input, \mathbf{u} , represents 334
 285 acceleration inputs to the system, we need no additional 335
 286 operations to transform the measured disturbances. That is, 336
 287 the bias vector is related to its compensation in the control 337
 288 law by an identity transform:

$$\mathbf{u}_{bc} = O_b \vec{\mathbf{B}} = \begin{pmatrix} -\mathbb{I}_{3 \times 3} \\ 0_{1 \times 3} \end{pmatrix} \vec{\mathbf{B}}. \quad (10)$$

289 **Feed-forward.** The final element of the control input is a 338
 290 feedforward signal that can be derived from a trajectory, $p(t)$, 339
 291 that is continuous and temporally smooth up to 3rd-order. For 340
 292 such paths, we have that $p(t), \dot{p}(t)$ as well as $\ddot{p}(t)$ are well- 341
 293 defined for all time t . The reference state for the feedback 342
 294 regulator, $\mathbf{x}_r \in \mathbb{R}^7$, is still composed only of $p(t)$, and $\dot{p}(t)$ 343
 295 (as well as heading).

296 Since multirotor systems are differentially flat, we know 344
 297 that by carefully selecting an output, $y_{df} = C_{df}\mathbf{x}$, we can 345
 298 express the system states as well as the system control inputs 346
 299 as functions of $y_{df}, \dot{y}_{df}, \ddot{y}_{df}$ and so on. In this case, we select 347
 300 only the translational position in three axes as the flat output, 348
 301 i.e., $C_{df} = \begin{pmatrix} \mathbb{I}_{3 \times 3} & 0_{3 \times 4} \end{pmatrix}$, and thus, $y_{df} = [p_n, p_e, p_d]^T$.

Again, since the control inputs to the system are accelera- 302
 tions, we can directly employ $\ddot{y}_{df} = \ddot{p}(t)$ as the feedforward 303
 control, such that, $\mathbf{u}_{ff} = \begin{pmatrix} \mathbb{I}_{3 \times 3} \\ 0_{1 \times 3} \end{pmatrix} \ddot{p}$.

Note that we do not design a feedforward component for 305
 the heading (yaw) control of the UAS. Since multirotors are 306
 typically invariant to yaw, and high accelerations in heading 307
 are less common in trajectories, we do not prioritize yaw 308
 agility in the outer-loop control in this work. However, if 309
 required, this can be incorporated by changing C_{df} and 310
 planning smooth trajectories for yaw. 311

The final control input from Eqn (8) is then,

$$\mathbf{u} = -K_{lqr}(\mathbf{x} - \mathbf{x}_r) + \begin{pmatrix} -\mathbb{I}_{3 \times 3} \\ 0_{1 \times 3} \end{pmatrix} \vec{\mathbf{B}} + \begin{pmatrix} \mathbb{I}_{3 \times 3} \\ 0_{1 \times 3} \end{pmatrix} \ddot{p}. \quad (11)$$

This represents the desired accelerations in three translational 312
 axes and one rotational axis (yaw) for the rigid body. As 313
 mentioned in Section III-A, using the total mass, m , the 314
 actual control input to the autopilot, $\mathbf{u}_{ap} = [\phi_d, \theta_d, \psi_d, T_d]^T$ 315
 can now be obtained by inverting Eqn (1). 316

IV. STUDIES 317

We now demonstrate the capabilities of the proposed 318
 architecture, along with the impact of its individual elements. 319
 The focus in these results is the ability of this pipeline 320
 to estimate and compensate for external disturbances, and 321
 execute dynamic trajectories with high precision in the field. 322
 We therefore select three illustrative scenarios that encompass 323
 a variety of our outdoor missions: hovering at a spot, flying 324
 in a circle, and executing a planned interception mission. 325
 For each of these, we will consider the time-sensitive tra- 326
 jectory tracking performance of our system, and its ability 327
 to reject external disturbances in all axes. For circles and 328
 more dynamic planned trajectories, our system benefits from 329
 incorporating a feedforward element. 330

Implementation Details 331

For the purposes of a fair and replicable evaluation, we 332
 implement the presented pipeline on a commercially available 333
 and fully open-source system. The hardware frame is an 334
 off-the-shelf DJI Flamewheel quadrotor with brushless DJI 335

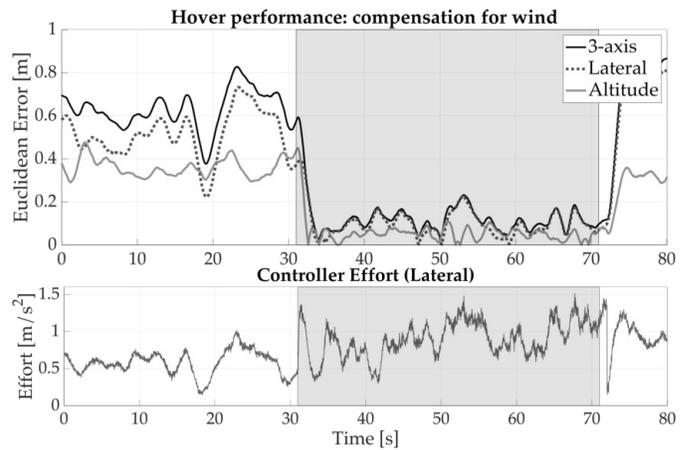


Fig. 3: Hover performance under wind speeds of up to 5.4 m/s. Wind compensation is active during the shaded region.

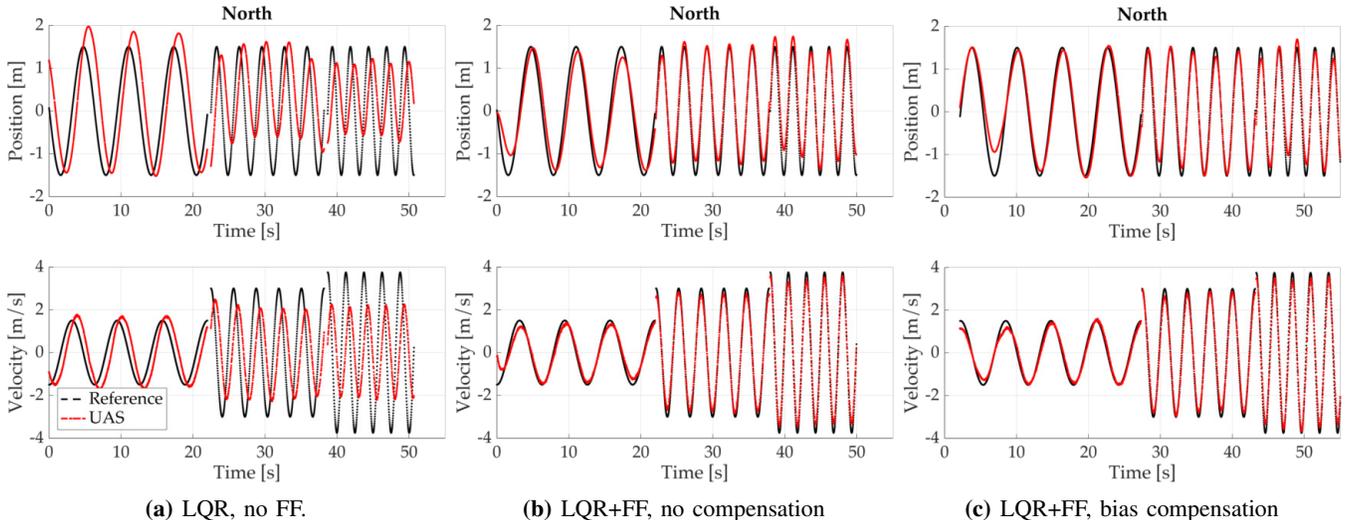


Fig. 4: Comparison of results in tracking circular trajectories of a fixed radius (1.5 m) and increasing angular rates with various elements of the pipeline enabled. (a) Naive LQR feedback with no feedforward and no bias compensation, (b) LQR with trajectory feedforward enabled, and, (c) LQR with trajectory feedforward and bias compensation from the full LQG system. Due to ambient wind, a steady offset can be observed in (b) which is corrected and centered in (c) by the bias estimation process. Ambient wind: 2–3 m/s N.

336 motor-ESC systems. The UAS measures ≈ 45 cm diagonally,
 337 weighs 1.2 kg with battery, and is capable of lifting more than
 338 an additional 1 kg. The autopilot is a commercial Pixhawk board
 339 running a fork of the open-source ArduCopter firmware. We equip
 340 the UAS with a u-blox ZED-F9P board that produces precise RTK-GPS
 341 data using standard GPS antennas at 5 Hz. The rest of the implementa-
 342 tion is all written in C/C++ over Robot Operating System (ROS) middle-
 343 ware stacks, and implemented entirely onboard on an Odroid XU4.
 344 This is made publicly available¹. The system model and
 345 feedback gains are developed on the complementary Freyja-
 346 Simulator². For instance, the gain matrix K can be obtained
 347 and validated in the simulator environment using MATLAB’s
 348 `place()` or `dlqr()` commands.

350 Our system architecture is easily adapted to several dif-
 351 ferent autopilot and UAS systems by only configuring the
 352 system parameters/scalars of the model. The pipeline pre-
 353 sented here has also been extensively employed and flight
 354 tested on Ascending Technologies’ autopilot and frames, in
 355 indoor motion-capture environments over wireless telemetry,
 356 and through other sources of state information such as an
 357 Intel RealSense T265 camera [22] and monocular vision
 358 pipelines both indoors and outdoors [24].

359 A. Hovering, Wind Resistance

360 In the first evaluation, we require the UAS to be positioned
 361 at a fixed 3D point in space under the presence of varying
 362 wind disturbances. Furthermore, to increase the estimation
 363 complexity, we specify a slightly higher mass in the system
 364 model (+0.1 kg), which results in a higher thrust than re-
 365 quired. These two combined effects are common in outdoor
 366 missions, specifically those which involve handling cargo.

367 Figure 3 shows the positioning Euclidean errors $\|\mathbf{x} - \mathbf{x}_r\|_2$
 368 from a fixed reference as a function of time. The average

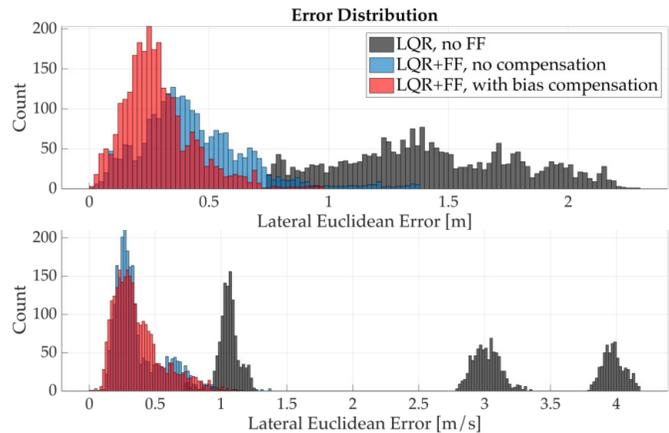


Fig. 5: Distribution of lateral trajectory tracking errors for position (top) and velocity (bottom) references. The three histograms represent data from the three columns in Figure 4.

369 wind speed during the flight is around 5 m/s. We switch the
 370 bias compensation on mid-flight (shaded region in figure) to
 371 capture its dynamics. We notice that the lateral (2D) and the
 372 3D errors are typically over 0.5 m when the compensation
 373 is inactive. When activated, the error rapidly diminishes to
 374 an average of ≈ 0.125 m in the shaded region. The estimator
 375 converges to its steady value within 2 s of activation, and also
 376 aids in reducing the vertical error due to an incorrect mass.

377 B. Circles

378 Next, we investigate the performance of the system over
 379 time-parameterized trajectories. As mentioned before, contin-
 380 uous and twice-differentiable paths can enable feed-forward
 381 elements in the controller, thereby aiding its temporal perfor-
 382 mance as well. Circles are well-suited for these tests, since
 383 the parametric cartesian forms are infinitely differentiable,
 384 and let us vary the translational speed targets (velocity norm
 385 in the lateral plane) in two axes.

¹github.com/unl-nimbus-lab/Freyja

²github.com/unl-nimbus-lab/Freyja-Simulator

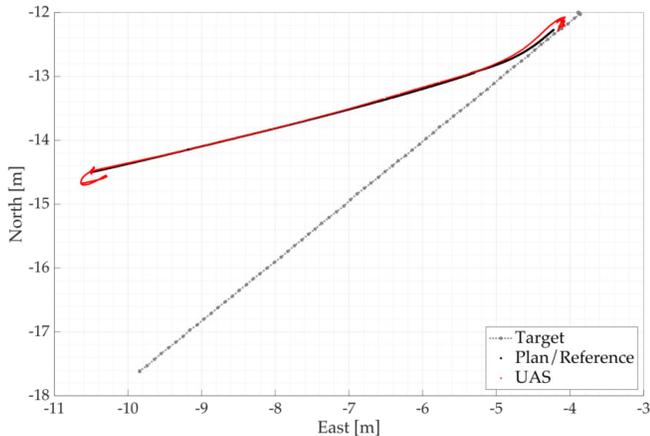


Fig. 6: Top-down (North-East) view of the docking experiment. The target and the UAS trajectories begin on the left.

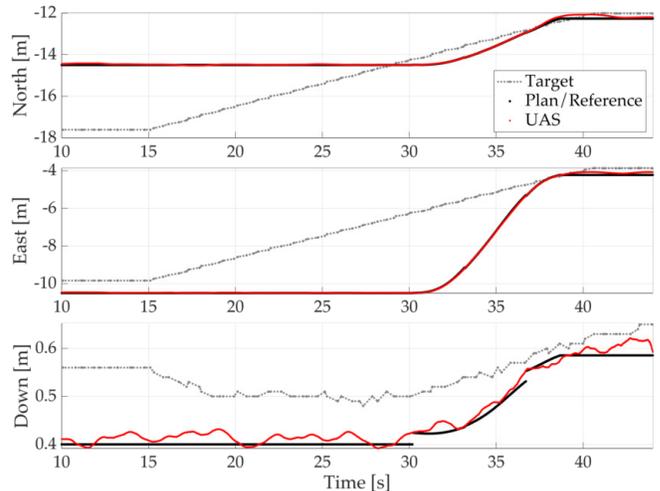


Fig. 7: Docking with a moving target by planning a smooth trajectory towards its projected (future) location.

In Figure 4 we show the North-component of the trajectories executed by the UAS outdoors in flying a reference circle of fixed radius and increasing angular rates. The vehicle is commanded peak lateral accelerations of almost 10 m/s^2 . We present results from three evaluations performed under a 2–3 m/s wind from North: using only position and velocity references in a classical feedback style (Fig. 4a), incorporating trajectory feedforward (Fig. 4b), and finally with the full LQG system (Fig. 4c). As expected, without the feedforward elements, the system lags behind in time with increasing angular rates. This behaviour is exacerbated when flying outdoors and external disturbances push the system away from a desired path. With feedforward enabled, we see that the tracking is more accurate and shows negligible lag. However, without compensating for external disturbances, the UAS trajectory has an upward shift (more prominent around 30 s). This is counteracted when bias compensation is enabled. Figure 1 shows a blended view of these aggressive trajectories with the UAS at a high lean angle.

Figure 5 also shows a histogram representation of the lateral position and velocity tracking errors seen in Figure 4. From the distribution, we see that the position errors (top) for a simple feedback system can fall between 0.75–2 m. When feed-forward and bias compensation from LQG are applied, the errors are reduced to less than 0.2 m. An interesting artifact of losing phase-tracking can also be seen in the velocity distributions when no feedforward is available.

C. Aerial Docking

Finally, we demonstrate an ultimate performance objective of the UAS in outdoor applications by tracking and predicting a future location of a moving target platform to dock with it in flight. In-flight docking is extremely challenging for multirotors due to a variety of safety and mechanical constraints. In this problem, we assume only that the target is moving in a predictable path (is not evasive), and that some intermittent observations of the target are available through its GPS data. To aid a fast recovery and accurate state estimation of the target, we also equip it with a passive fiducial marker that can be observed by an onboard camera in close approaches ($< 2\text{--}3 \text{ m}$). The full pipeline presented here is employed for UAS

control, but the relative pose estimation for the target over a horizon is accomplished by fusing these complementary modalities of information. This lets us plan (and replan) a smooth and efficient trajectory towards this projected final location, and engage a mechanical actuator to dock. Detailed and in-depth evaluations under various outdoor scenarios are available [21]; here we focus on path following capabilities.

Figure 6 shows the top-down (North-East) view of the target’s path, and the interception plan generated and executed by the UAS. In this particular instance, the target is a zipline system that moves in a straight-line in the lateral plane, but affects a parabolic sag in the vertical axis. We see that the planned path meets the target’s path at the highlighted region, and that the UAS also executes it correctly.

A temporal view of the same experiment is shown in Figure 7 for all three axes. The actual successful docking occurs at around the 38 s mark, and the UAS starts its path around 30 s (prior to that, observations are being collected to estimate the target’s trajectory). Once again, we see that the UAS follows the reference trajectory precisely in space and time, which is crucial for a planned time-critical missions. Also note that the scale on ‘Down’ axis has more than 10x magnification; the overall 3D accuracy in hover is $\approx 4 \text{ cm}$.

V. CONCLUSIONS & REMARKS

We have presented a complete framework, *Freyja*, that sequentially addresses each aspect of a multirotor flight in real and challenging outdoor environments. The full open-source pipeline is structurally modular, incorporates several optimal methods from the literature to enable precise maneuvering in agile flight maneuvers, and is amenable to extension as the state of the art progresses. For instance, while *Freyja*’s state-space representation of Eqn 1 for the controller enables easy integration of 3D path planners, it currently precludes acrobatic trajectories in the rotational space (such as flips and inverted flight). Our extensive field results demonstrate the capabilities of the system in rejecting environmental disturbances and precisely executing time-critical trajectories.

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